

AN OSCILLATOR USING THE NAND SCHMITT-TRIGGER

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A simple method to generate rectangular waves is with NAND Schmitt-trigger the gate.

The oscillator here presented uses only one of the four gates existent in the commercial integrated circuit CD4093 or its equivalent circuits.

The advantage of the circuit shown here is that it is possible to control the frequency and duty-cycle in an almost independent manner.

In the diagram we see that the capacitor charges by the 'high' output through the resistor R2 and the part Rd of R1 to the right of its center, as its left part is short-circuited by the left diode.

As the threshold of the IC is reached, this produces an output 'low' that discharges the capacitor through R2 and the left part Re of R1, as its right part is short-circuited by the right diode. So, the charge time constant depends on $R2 + Rd$ and the discharge time constant depends on $R2 + Re$. As the period of the oscillation is the sum of those two time constants, it depends on $R2 + Re + R2 + Rd$. But $Re + Rd = R1$, and the period depends only on $2 \times R2 + R1$, that is, independent of the value of the control R1.

Thus, R1 controls the ration between the charge and discharge times, that is, the duty-cycle, and the position of R2 controls the frequency. The maximum frequency is given for $R2 = 0$, that is, it is limited by value of R1 (of course, also by the characteristics of the IC).

The above calculi consider the diodes as ideal ones; therefore, germanium diodes are more suitable for greater independence of both controls.

The charge time T_c corresponds to the voltage increase from V_{min} to V_{max} . During the charge the capacitor voltage is given by:

$V_c = A + B \cdot \exp(-t/\tau_c)$ where τ_c = charge time constant; as, for $t = 0$, $V_c = V_{min}$ and for $t \rightarrow \infty$, $V_c \rightarrow V_{max}$, we have:

$V_c = V_{cc} + (V_{min} - V_{cc}) \cdot \exp(-t/\tau_c)$; as, for $t = T_c$, $V_c = V_{max}$, we have:

$$V_{max} = V_{cc} + (V_{min} - V_{cc}) \cdot \exp(-T_c/\tau_c) \text{ or } T_c = \tau_c \cdot \ln [(V_{cc} - V_{min}) / (V_{cc} - V_{max})]$$

\ln stands for the natural logarithm.

The threshold 'up' and 'down' voltages **V_{max}** and **V_{min}** for the used IC are respectively 60% and 40% (typical) of the supply voltage **V_{cc}** (this varies with temperature and with **V_{cc}** itself). So, we can write for the charge time:

$$\mathbf{T_c = 0.4 \cdot \tau_c \quad (1)}$$

The discharge time **T_d** corresponds to the decrease from **V_{max}** to **V_{min}**. During the discharge the capacitor voltage is given by:

V_c = a + b . exp(-t/τ_d), where **τ_d** = discharge time constant; as, for **t = 0**, **V_c = V_{max}**, and for **t → ∞**, **V_c → V_{min}**, we have:

V_c = V_{max} . exp(-t/τ_d); as, for **t = T_d**, **V_c = V_{min}**, we have:

V_{min} = V_{max} . exp(-T_d/τ_d) or **T_d = τ_d . ln (V_{max} / V_{min})**. With their values, we can write for the discharge time:

$$\mathbf{T_d = 0.4 \cdot \tau_d \quad (2)}$$

As (1) e (2) have the same constant 0.4, we have for the total period **T**:

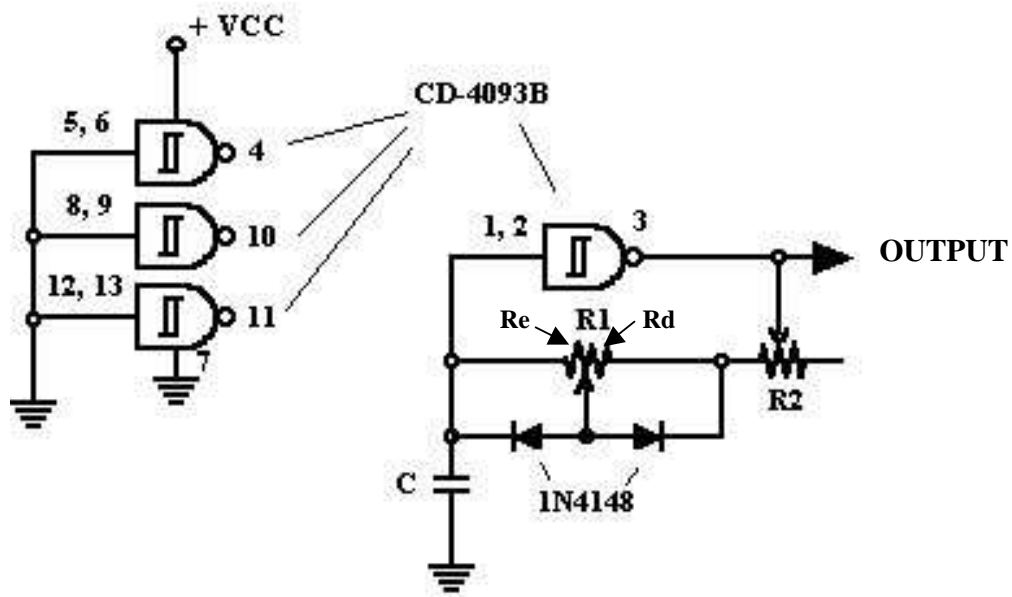
$$\mathbf{T = 0.4 \cdot (\tau_c + \tau_d) \quad (3)}$$

But **τ_c = R₂ + R_d** and **τ_d = R₂ + R_e** and so we have, with **R_d + R_e = R₁**:

T = 0.4 . (2R₂ = R₁) or the frequency **F**:

$$\mathbf{F = 1 / [0.4 \cdot (2R_2 + R_1)]}$$

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The symmetry is controlled by R1
The frequency is controlled by R2

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